

# Description

BP3108 is a primary side regulation (PSR) controller with triac dimming designed for high brightness LED driver. It's extremely suitable to be used in flyback converter operating in worldwide voltage range.

BP3108 works in discontinuous inductive current mode. The patented control method dramatically saves the system cost and size by eliminating all secondary feedback circuitry and loop control compensation circuitry. This control method also achieves excellent line and load regulation.

BP3108 integrates the triac dimming control circuit internally.Few external components are required to achieve wide triac dimming range and good dimming effect without visible flicker.

BP3108 integrates high precision current sense circuitry, which achieves  $\pm 3\%$  LED current accuracy.

BP3108 offers rich protection features including DED short/open circuit protection, over-temperature protection, under-voltage lockout, VCC over voltage protection.

BP3108 is assembled in SOP-8 package.

**Typical Application** 

# Features

- Triac Dimming
- PSR constant current (CC) control, no secondary feedback circuit required
- $\pm 3\%$  LED current accuracy
- Ultra low operating current to improve efficiency
- Low power consumption on FB resistor network
- ◆ Worldwide voltage range ∡
- ◆ LED short/open circuit protection
- $V_{CC}$  over voltage protection
- Current sense resistor open circuit protection
- Over temperature protection
- No external loop compensation required

# Applications

- Triac dimming LED lighting
- GU10/E27 LED bulb, spot light
- ◆ LED PAR30/PAR38, down light
- LED Tube
- Other LED lighting

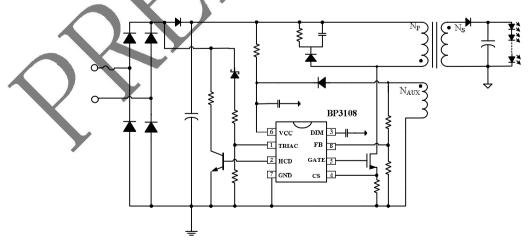


Figure 1 typical application circuit



# **Pin Configuration**

Pin

| TRIAC   |          |   |  |  |  |
|---------|----------|---|--|--|--|
| Pin No. | Pin Name | Description   |  |  |  |
| 1       | TRIAC    | Triac dimmng voltage detect, PWM dimming signal input   |  |  |  |
| 2       | HCD      | Dimming BJT driver  |  |  |  |
| 3       | DIM      | Analog dimming Connect 1uF capacitor to GND while triac dimming.  |  |  |  |
| 4       | CS       | The current sense pin used to sense the MOSFET current by means of an external sense resistor. The resistor is connected between CS and GND |  |  |  |
| 5       | GATE     | The output gate driver for an external N-channel power MOSFET.  |  |  |  |
| 6       | VCC      | Power supply  |  |  |  |
| 7       | GND      | Signal and Power ground   |  |  |  |
| 8       | FB.      | The voltage feedback from auxiliary winding.  |  |  |  |

# **Ordering Information**

| Order Number | Package | Operating<br>Temperature | Packing Method            | Marking                 |
|--------------|---------|--------------------------|---------------------------|-------------------------|
| BP3108ESO8   | SOP8    | -40 °C to 85 °C          | Tape and Reel 2,500 units | BP3108<br>xxxxxx<br>xxx |
| Y            |         |                          |                           |                         |



| Symbol            | Parameters                                   | Value      | Unit |
|-------------------|--|------------|------|
| V <sub>CC</sub>   | Power supply                                 | -0.3~25    | V    |
| FB                | The voltage feedback from auxiliary winding. | -0.3~6     | V    |
| CS                | Current sense                                | -0.3~6     | V    |
| DIM               | Analog dimming                               | -0.3~6     | V    |
| TRIAC             | Triac dimmng voltage detect                  | -0.3~6     | V    |
| HCD               | Dimming BJT driver                           | -0.3~6     | v    |
| Gate              | Gate driver of external NMOS                 | -0.3~25    | V    |
| P <sub>DMAX</sub> | Power dissipation (note2)                    | 0.3        | W    |
| P <sub>TR</sub>   | Thermal resistance, SOP-8 $(\theta_{JA})$    | 145        | °C/W |
| TJ                | Operating temperature                        | -40 to 150 | °C   |
| T <sub>STG</sub>  | Storage temperature                          | -55 to 150 | °C   |
|                   | ESD (note3)                                  | 2          | kV   |

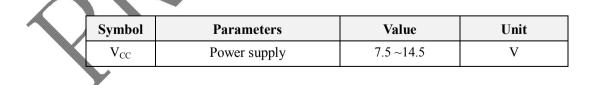
#### Absolute Maximum Ratings (note1)

Note 1: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. Under "recommended operating conditions" the device operation is assured, but some particular parameter may not be achieved. The electrical characteristics table defines the operation range of the device, the electrical characteristics is assured on DC and AC voltage by test program. For the parameters without minimum and maximum value in the EC table, the typical value defines the operation range, the accuracy is not guaranteed by spec.

**Note 2:** The maximum power dissipation decrease if temperature rise, it is decided by  $T_{JMAX}$ ,  $\theta_{JA}$ , and environment temperature  $(T_A)$ . The maximum power dissipation is the lower one between  $P_{DMAX} = (T_{MAX} - T_A)/\theta_{JA}$  and the number listed in the maximum table.

Note 3: Human Body mode, 100pF capacitor discharge on  $1.5k\Omega$  resistor

# **Recommended Operation Condition**





| Parameter                         | Conditions   | Min.  | Тур.   | Max.   | Unit   |
|-----------------------------------|--|---|--|--|--|
| ction                             |  |   |  |  |  |
| Input voltage                     |  | 6.5   | 12   | 16   | V  |
| Start-up voltage                  | V <sub>CC</sub> rising   | 13  | 14   | 15   | V  |
| Under-voltage lockout threshold   |  |   | 6.5  | 1  | V  |
| Over voltage protection threshold |  |   | 16   |  | V  |
| V <sub>CC</sub> clamp voltage     | Ivcc=10mA  |   | 19   |  | V  |
| etion                             |  |   |  | 2  | <b>X</b>   |
| Current sense threshold           |  | 495   | 500  | 505  | mV   |
| LEB time                          |  |   | 500  |  | ns   |
| Switch off delay                  |  |   | 200  |  | ns   |
| it section                        |  |   |  |  |  |
| Start up current                  | $V_{CC} = V_{CC-TH} - 1V$  |   | 25   | 50   | uA   |
| Typical operation current         | F <sub>OP</sub> =40kHz   | $\mathbf{\mathbf{Y}}$   | 1.5  |  | mA   |
|                                   |  |   |  |  |  |
| FB voltage threshold              |  |   | 1  |  | V  |
| FB clamp voltage                  | I <sub>FB</sub> =2µA   |   | 1.5  |  | V  |
| Minimum demagnetization time      | $\mathbf{\mathbf{Y}}$  |   | 4.2  |  | uS   |
| Line compensation ratio           | (note 6)   |   | 1.1  |  | mV/uA  |
| cycle                             |  |   |  |  |  |
| Maximum Duty Cycle                |  |   | 42   |  | %  |
| ection                            |  |   |  |  |  |
| TRIAC pull up resistor            |  |   | 2  |  | MΩ   |
| HCD switch off resistor           |  |   | 12   |  | Ω  |
| HCD sourcing current              |  |   | 500  |  | uA   |
| TRIAC current                     |  |   | 5  |  | uA   |
| TRIAC votaget dectect threhold    |  |   | 1.25   |  | V  |
| section                           |  |   |  |  | L  |
| Analog dimming low threshold      |  |   | 0.8  |  | V  |
| Analog dimming high threshold     |  |   | 3.5  |  | V  |
| 0 0 0                             |  |   |  |  |  |
|                                   | ction Input voltage Start-up voltage Under-voltage lockout threshold Over voltage protection threshold Vcc clamp voltage ction Current sense threshold LEB time Switch off delay tsection Start up current Typical operation current FB voltage threshold FB clamp voltage FB clamp voltage Minimum demagnetization time Line compensation ratio cycle Maximum Duty Cycle Cycle HCD switch off resistor HCD switch off resisto | ctionInput voltageVcc risingStart-up voltage lockout thresholdVcc risingUnder-voltage protection thresholdIvcc=10mAVcc clamp voltageIvcc=10mAVcc clamp voltageIvcc=10mAtionIvcc=10mACurrent sense thresholdILEB timeIvcc=10mASwitch off delayVcc vcc-rn+1Vt sectionVcc=Vcc-rn+1VTypical operation currentVcg=Vcc-rn+1VFB voltage thresholdFop=400HzFB clamp voltageIm=2uAMinimum demagnetization timeIm=2uALine compensation ratio(note 6)cycleIm=2uAMaximumDuty CycleIm=2uAHCD sourcing currentIm=2uAHCD sourcing currentIm=2uATRIAC pultup resistorIm=2uAHCD sourcing currentIm=2uATRIAC currentIm=2uATRIAC votaget dectect threholdIm=2uAsectionIm=2uAAnalog dimming low thresholdIm=2uAImage dimming low thresholdIm=2uAImage dimming low thresholdIm=2uAImage dimming low thresholdImage dimming low threshold | ction6.5Input voltage6.5Start-up voltageVccrisingInder-voltage lockout thresholdImage of the sholdOver voltage protection thresholdImage of the sholdVcc clamp voltageImage of the sholdVcc clamp voltageImage of the sholdCurrent sense thresholdImage of the sholdSwitch off delayImage of the sholdStart up currentVcc=Mcc-m-1WStart up currentVcc=Mcc-m-1WTypical operation currentFor=40kHzFB voltage thresholdImage of the sholdFB clamp voltageImage of the sholdFB clamp voltageImage of the sholdInimum demagnetization timeImage of the sholdInimum demagnetization timeImage of the sholdFullImage of the sholdFIAC pull up resistorImage of the sholdHCD switch off resistorImage of the shold | ction $1.1$ Input voltage $6.5$ $12$ Start-up voltage $V_{cc}$ rising $13$ $14$ Under-voltage lockout threshold $1.6$ $6.5$ Over voltage protection threshold $1.6$ $16$ $V_{cc}$ clamp voltage $Ivce=10mA$ $19$ ction $Ivce=10mA$ $19$ Current sense threshold $495$ $500$ LEB time $495$ $500$ Switch off delay $200$ ts ection $V_{cc}=V_{cc-mt}-1V$ $25$ Typical operation current $F_{op}=40kHz$ $1.5$ FB voltage threshold $1$ $1$ FB clamp voltage $hn=2pA$ $1.5$ Minimum demagnetization time $4.2$ $4.2$ Line compensation ratio(note 6) $1.1$ typical operation current $flack pullup resistor2flack pullup resistor242texton12500tractor12HCD switch off resistor12HCD switch off resistor12HCD switch off resistor500TRIAC votaget dectect threhold50TRIAC votaget dectect threhold50TRIAC votaget dectect threhold50TRIAC system colspan="4">SuperiorAnalog dimming low threshold0.8$ | ctionInput voltage $6.5$ 1216Start-up voltage $V_{cc}$ rising131415Under-voltage lockout threshold $6.5$ $12$ $16$ Over voltage protection threshold $16$ $6.5$ $7$ Over voltage protection threshold $16$ $16$ $7$ Vcc clamp voltageIvec=10mA $19$ $7$ Current sense threshold $495$ $500$ $505$ LEB time $500$ $505$ $505$ LEB time $200$ $7$ Start up current $V_{cc}$ $k_{ccm}$ $1V$ $25$ $50$ Typical operation current $F_{0r}$ $404tiz$ $1.5$ $7$ FB voltage threshold $1$ $1$ $1$ FB clamp voltage $u_{c} = 2pA$ $1.5$ $7$ Minimun demagnetization time $u_{c} = 2pA$ $1.5$ $7$ THAC pult up resistor $10$ $12$ $12$ true $12$ $12$ $12$ $12$ true $12$ $500$ $12$ $12$ true $500$ $12$ $125$ $125$ true $12$ $125$ $125$ $125$ true $125$ $500$ $125$ $125$ true $125$ $500$ $125$ $125$ true $125$ $500$ $125$ $125$ true |

# Electrical Characteristics (V<sub>cc</sub>=12V, T<sub>A</sub>=25 °C unless otherwise stated)(note 4,5)



# **BP3108** *PSR LED Controller With Triac Dimming*

| V <sub>DIMPULL-UP</sub> | Dim pull resistor           |   | 150 |              | KΩ |
|-------------------------|-----------------------------|---|-----|--------------|----|
| Over temperat           | ure protection              |   |     |              |    |
| T <sub>SD</sub>         | Thermal shutdown threshold  |   | 160 |              | °C |
| T <sub>SD-HYS</sub>     | Thermal shutdown hysteresis |   | 30  |              | °C |
| Gate Driver see         | tion                        | • | •   | •            |    |
| I <sub>SOURCE</sub>     | Gate sourcing current       |   | 100 | ~            | mA |
| R <sub>DSOFF</sub>      | Gate switch off resistor    |   | 30  | $\checkmark$ | Ω  |

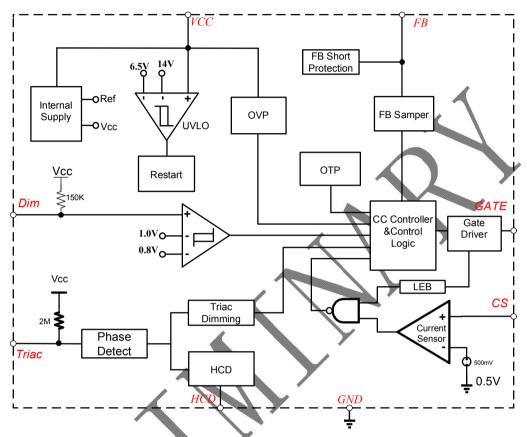
*Note4* : production testing of the chip is performed at 25  $^{\circ}$  C.

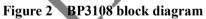
Note5: the maximum and minimum parameters specified are guaranteed by test, the typical value are guaranteed by design, characterization and statistical analysis

Note6: refer to application information



#### **Application Information**





BP3108 is a high performance AC/DC controller with triac dimming specially designed for LED lighting. The patented triac dimming control and CC control method dramatically saves the system cost and size by eliminating all secondary feedback circuitry and loop control compensation circuitry. Few external components are required to achieve wide triac dimming range and good dimming effect without visible flicker.

#### 1 Start up

The start-up current is designed to be low, only 25uA start-up current is required for the device. After the system is powered on, the Vcc capacitor is charged through the start-up resistor, once the Vcc voltage

reaches the start-up threshold (14V typically), the chip start to operate. During normal operation, The Vcc is supplied via the auxiliary winding.

#### 2 Constant current control

Cycle-by-Cycle current limiting is offered .The CS is connected to non-inverting inputs of the current sense comparator. The inverting terminal of comparator is tied to an internal 500mV reference. The comparator is constantly comparing the CS voltage to the internal 500mV.Once the output of comparator goes high, the GATE output goes low. The primary peak current is given by:

$$I_P = 500(\text{ mV}) / R_{CS}(mA)$$

The outputs of the comparators also include a typical 500ns (typ.) blanking time which prevents spurious



turn-offs of the external MOSFET due to the turn-on spike normally.

The LED current is decided by the internal constant current control, which is given by:

$$I_{OUT} = \frac{I_P}{4} \times \frac{N_P}{N_S}$$

Np is turns of transformer primary winding, Ns is turns of transformer secondary winding, Ip is the peak current of the primary side

#### **3** Feedback network

The chip senses the output voltage and output current through the feedback network. The FB sense threshold is 1V. We recommend to set the feedback upper resistor to  $300 \text{K}\Omega \sim 750 \text{K}\Omega$  to improve system efficiency.

The FB upper resistor is also used to adjust the Vcs threshold to compensation the input voltage change. It is given by:

$$\Delta V_{CS} = -1.1 \times 10^6 * \frac{V_{AUX}}{R_{FBH}} (mV)$$

 $R_{FBH}$  is the upper resistor of the feedback network

## 4 Gate drive

The driver structure of BP3108 is optimized. Too strong drive current results in bad EMI while too weak driver cause big switch loss of power MOS. A good trade-off is achieved in the internal driver with suitable strength.

## **5** Operation switching frequency

The system work in DCM mode, the maximum duty cycle is 42%; Normally the system frequency is recommended to be set between 40KHz-48KHz to get better EMI result. The maximum operating frequency should be lower than 80KHz, while the minimum system frequency should be higher than 20KHz. The operation frequency is given by:

$$f = \frac{N_P^2 * V_{LED}}{8 * N_S^2 * L_P * I_{LED}}$$

Lp is inductance of primary winding.

The chip limits the maximum system frequency and the minimum system frequency to ensure the system stability.

### **6** Protection control

Good power supply system reliability is achieved with its rich protection features.

Once the Vcc voltage is higher than 16V (typ.), the device shut down the MOSFET and goes into auto-start mode until the over-voltage is removed. Another 19V clamp circuitry is offered to prevent the chip from damage on abnormal situation.

The thermal shutdown circuitry senses the die temperature. The threshold is set at 160°C typical with a 30°C hysteresis. When the die temperature rises above this threshold (160 °C) the power MOSFET is disabled and remains disabled until the die temperature falls by 30 °C, at which point the MOSFET is re-enabled.

Once LED short circuit or LED open circuit is detected, the system enters into low standby mode in which the system consumes very low power and the output status is monitored continuously. The chip goes back to normal operation automatically after the abnormal situation is removed.

## 7 Triac dimming

BP3108 integrates the triac dimming control circuit internally .The patented triac dimming control and phase detect technology is compatiable to most triac dimmer.

## **8 PCB** Layout

The following rules should be followed in BP3108 PCB layout.



#### **Bypass capacitor**

The Vcc bypass capacitor should be as close to the Vcc pin as possible.

#### Current sense (CS) resistor

The CS resistor ground should be as close to the chip ground as possible to reduce the ground noise coupled, which improves the quality of the sensing signal.

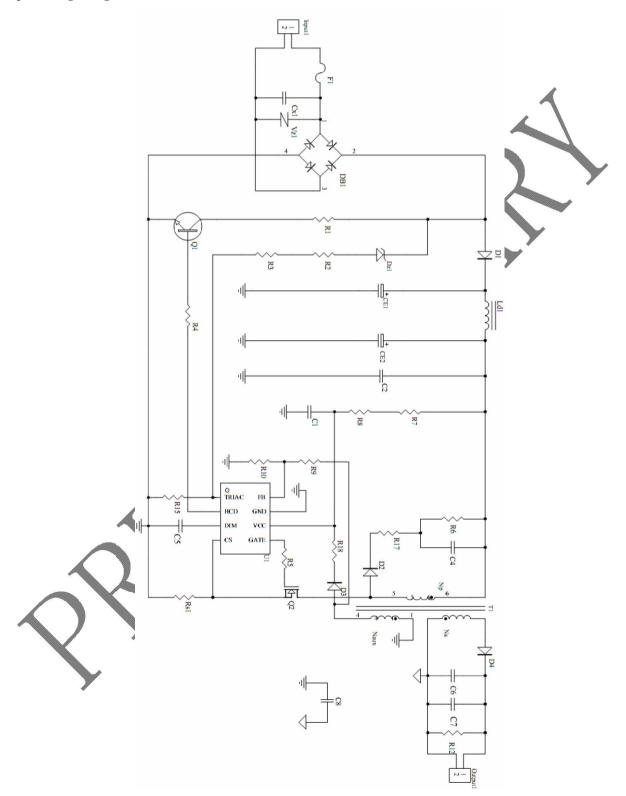
#### Current path area

The area of current path should be as small as possible to reduce EMI radiation, such as the primary current loop from the primary inductor to the MOSFET and the secondary current loop from the secondary inductor to the output rectifier.



# **Application Example**

Input voltage range: 85V-264V with triac dimmin





#### **Package Information**

