

Introduction (General Description)

The EC5565 is a 4+1 channel voltage buffers that buffers reference voltage for gamma correction in a thin film transistor liquid crystal display (TFT LCD). This device incorporates a Vcom amplifier circuits, four rail to rail buffer amplifier circuits.

The EC5565 is available in a space saving 14-pin TSSOP package and, QFN 16L package. and the operating temperature is from -20°C to +85°C.

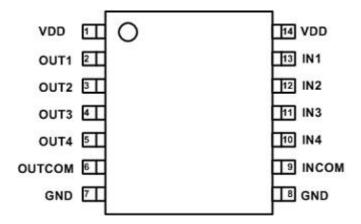
Features

- Wide supply voltage range 6.5V ~ 18V
- Rail-to-Rail output swing (The highest two stage & lowest two stage)
- High slew rate 1V/ μ s
- GBWP 1 MHz
- 2 MHz -3dB Bandwidth
- Large Vcom Drive Current: ±100mA(Max.)
- Ultra-small Package TSSOP-14 and QFN 16L.

Applications

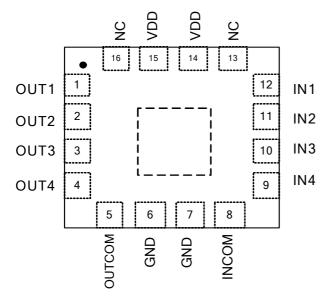
TFT-LCD Reference Driver

Pin Assignment For TSSOP-14





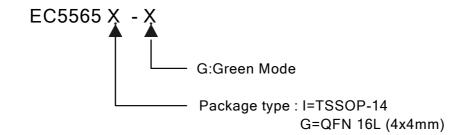
Pin Assignment For QFN 16L (Top View)



Ordering Information

PART NUMBER	MAKING	PACKAGE
EC5565I	AS05	TSSOP-14
EC5565I-G	AS05-G	GREEN Mode, TSSOP-14
EC5565G-G	AS05-G	GREEN Mode, QFN 16L (4x4 mm)

Ordering Information





Absolute maximum ratings (TA = 25 °C)

Values beyond absolute maximum ratings may cause permanent damage to the device. These are stress ratings only; functional device operation is not implied. Exposure to AMR conditions for extended periods may affect device reliability.

Parameter	Symbol	Value	Unit
Supply Voltage between $V_{\mathbb{S}^+}$ and $V_{\mathbb{S}^-}$	Vs	+18	V
Input Voltage	Vin	V _{s-} -0.5	V
(For rail to rail)	VIII	V _{s+} +0.5	V
Maximum Output Current (1 ~ 4 Buffers)	lout	±30	mA
Maximum Output Current (Com Buffer)	lout(com)	±100	mA
Maximum Junction Temperature	TJ	+125	°C
Storage Temperature Range	TSTG	-65 to +150	°C
Operating Temperature Range	TOP	-20 to +85	°C
Lead temperature	Tlead	260	°C
ESD Voltage	VESD	2	KV

Important Note:

All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$



Electrical Characteristics (Typical Performance Characteristics)

 V_{S+} = +5V, V_{S-} = -5V, R_L = 10k Ω and C_L = 10pF to 0V, T_A = 25°C unless otherwise specified.

Parameter	Description	Condition	Min	Тур	Max	Units
Input Characteri	stics				1	
Vos	Input Offset Voltage	V _{CM} = 0V		2	12	mV
TCVos	Average Offset Voltage Drift	[1]		5		μV/°C
I _B	Input Bias Current	V _{CM} = 0V		2	50	nA
R _{IN}	Input Impedance			1		G
C _{IN}	Input Capacitance			1.35		pF
Output Characte	ristics		•		•	•
V _{OL}	Output Swing Low	I _L = -5mA (O1, O2, O3, O4 rail-to-rail Buffers)		-4.92	-4.85	V
V _{OH}	Output Swing High	I _L = 5mA (O1, O2, O3, O4 rail-to-rail Buffers)	4.85	4.92		V
I _{SC}	Short Circuit Current	(Out1 ~ Out4 Buffers)		±120		mA
I _{OUT}	Output Current	(Out1 ~ Out4 Buffers)		±30		mA
I _{SC(Com)}	Short Circuit Current	(Com Buffer)		±300		mA
I _{OUT(Com)}	Output Current	(Com Buffer)		±100		mA
Power Supply Pe	erformance					
PSRR	Power Supply Rejection Ratio	V _S is moved from ±3.25V to ±7.75V	60	80		dB
Is	Supply Current (Per Amplifier)	No Load (Out1 ~ Out4 Buffers)		500	750	μΑ
I _{S(Com)}	Supply Current	(Com Buffer)		2.5		mA
Dynamic Perforn	nance					
SR	Slew Rate [2]	-4.0V≦V _{OUT} ≦4.0V, 20% to 80%		1		V/µs
ts	Settling to +0.1% (AV = +1)	(AV = +1), V _O =2V Step		5		μs
BW	-3dB Bandwidth	RL = 10KΩ, CL = 10PF		2		MHz
PM	Phase Margin	RL = $10K\Omega$, CL = $10PF$		60		Degree
CS	Channel Separation	f = 1 MHz		75		dB

^{1.} Measured over operating temperature range

^{2.} Slew rate is measured on rising and falling edges



Typical Performance Characteristics

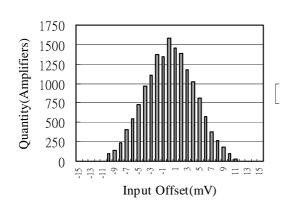


Figure (a) Input Offset Voltage Distribution

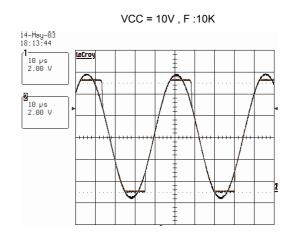


Figure (b) Input beyond the rails

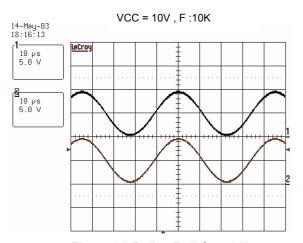


Figure (c) Rail to Rail Capability

VCC = 10V , F :1K

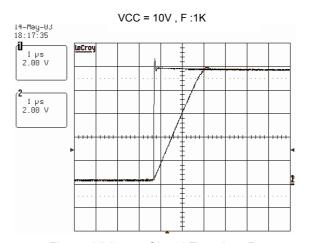


Figure (d) Large Signal Transient Response

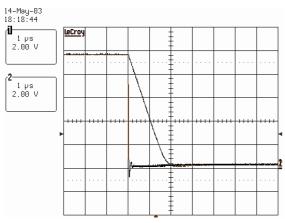


Figure (e) Large Signal Transient Response

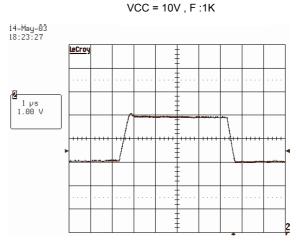


Figure (f) Small Signal Transient Response



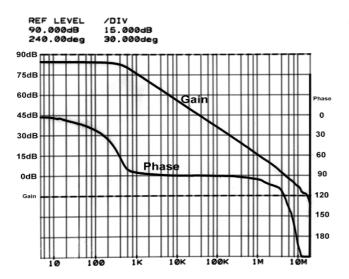


Figure (g) Open Loop Gain & Phase vs. Frequency

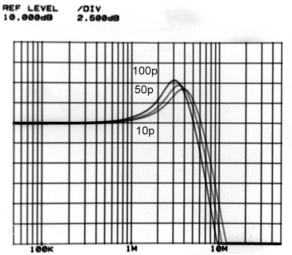


Figure (h) Frequency Response for Various C_L



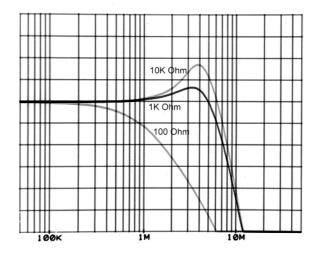


Figure (i) Frequency Response for Various R_L



Applications Information

Product Description

The EC5565 rail-to-rail 5 channels amplifier is built on an advanced high voltage CMOS process. It's beyond rails input capability and full swing of output range made itself an ideal amplifier for use in a wide range of general-purpose applications. The features of 1µS high slew rate, fast settling time, 2MHz of GBWP as well as high output driving capability have proven the EC5565 a good voltage reference buffer in TFT-LCD for grayscale reference applications. High phase margin and extremely low power consumption (500µA per amplifier) make the EC5565 ideal for Connected in voltage follower mode for low power high drive applications

Supply Voltage, Input Range and Output Swing

The EC5565 can be operated with a single nominal wide supply voltage ranging from 6.5V to 18V with stable performance over operating temperatures of -20 °C to +85 °C.

With 500mV greater than rail-to-rail input common mode voltage range and 75dB of Common Mode Rejection Ratio, the EC5565 allows a wide range sensing among many applications without having any concerns over exceeding the range and no compromise in accuracy. The output swings of the EC5565 typically extend to within 80mV of positive and negative supply rails with load currents of 5mA. The output voltage swing can be even closer to the supply rails by merely decreasing the load current. Figure 1 shows the input and output waveforms for the device in the unity-gain configuration. The amplifier is operated under ±5V supply with a 10k, load connected to GND. The input is a 10Vp-p sinusoid. An approximately 9.985 Vp-p of output voltage swing can be easily achieved.

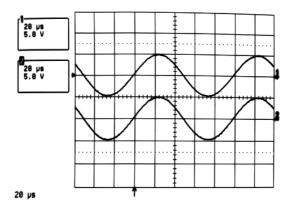


Figure 1. Operation with Rail-to-Rail Input and Output

Output Short Circuit Current Limit

A +/-120mA short circuit current will be limited by the EC5565 if the output is directly shorted to the positive or the negative supply. For an indefinitely output short circuit, the power dissipation could easily increase such that the device may be damaged. The internal metal interconnections are well designed to prevent the output continuous current from exceeding +/-30 mA such that the maximum reliability can be well maintained.



Output Phase Reversal

The EC5565 is designed to prevent its output from being phase reversal as long as the input voltage is limited from V_{S-} -0.5V to V_{S+} +0.5V. Figure 2 shows a photo of the device output with its input voltage driven beyond the supply rails. Although the phase of the device's output will not be reversed, the input's over-voltage should be avoided. An improper input voltage exceeds supply range by more than 0.6V may result in an over stress damage.

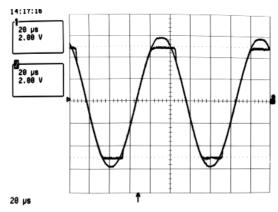


Figure 2. Operation with Beyond-the Rails Input

Power Dissipation

The EC5565 is designed for maximum output current capability. Even though momentary output shorted to ground causes little damage to the device.

For the high drive amplifier EC5565, it is possible to exceed the 'absolute-maximum junction temperature' under certain load current conditions. Therefore, it is important to calculate the maximum junction temperature for the application to determine if load conditions need to be modified for the amplifier to remain in the safe operating area. The maximum power dissipation allowed in a package is determined according to:

$$P_{\text{Dmax}} = \frac{T_{\text{Jmax}} - T_{\text{Amax}}}{\Theta_{\text{JA}}}$$

Where:

T_{Jmax} = Maximum Junction Temperature

T_{Amax} = Maximum Ambient Temperature

 Θ_{JA} = Thermal Resistance of the Package

 P_{Dmax} = Maximum Power Dissipation in the Package.

The maximum power dissipation actually produced by an IC is the total quiescent supply current times the total power supply voltage, plus the power in the IC due to the loads, or:

$$P_{Dmax} = \sum_{i} [V_S * I_{Smax} + (V_{S+} - V_O) * I_L]$$

When sourcing, and

$$P_{Dmax} = \sum_{i} [V_{S} * I_{Smax} + (V_{O} - V_{S}) * I_{L}]$$

When sinking.



Where:

i = 1 to 4

V_s = Total Supply Voltage

I_{Smax} = Maximum Supply Current Per Amplifier

 V_0 = Maximum Output Voltage of the Application

IL= Load current

 R_L = Load Resistance = $(V_{S+} - V_O)/I_L = (V_O - V_{S-})/I_L$

A calculation for R_L to prevent device from overheat can be easily solved by setting the two P_{Dmax} equations equal to each other.

Pin Count	Θja (°C/W)	Θjc (°C/W)
TSSOP-14	100	17
QFN 16L	43	8

Driving Capacitive Loads

The EC5565 is designed to drive a wide range of capacitive loads. In addition, the output current handling capability of the device allows for good slewing characteristics even with large capacitive loads. The combination of these features make the EC5565 ideally for applications such as TFT LCD panel grayscale reference voltage buffers, ADC input amplifiers, etc.

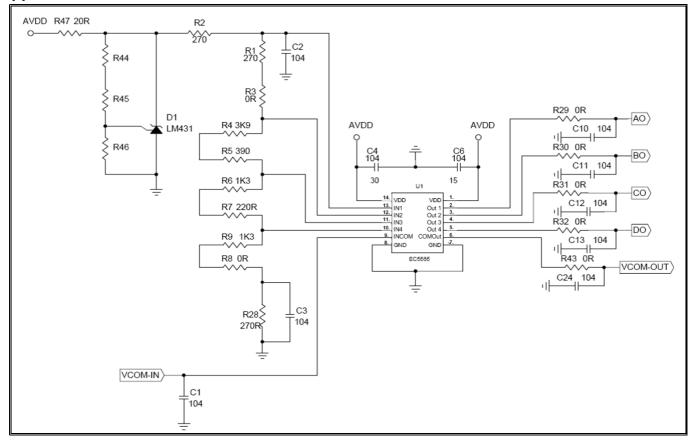
As load capacitance increases, however, the -3dB bandwidth of the device will decrease and the peaking increase. The amplifiers drive 10pF loads in parallel with10K Ω . with just 1.5dB of peaking, and 100pF with 6.4dB of peaking. If less peaking is desired in these applications, a small series resistor (usually between 5Ω and 50Ω) can be placed in series with the output. However, this will obviously reduce the gain slightly. Another method of reducing peaking is to add a "snubber" circuit at the output. A snubber is a shunt load consisting of a resistor in series with a capacitor. Values of 150Ω and 10nF are typical. The advantage of a snubber is that it improves the settling and overshooting performance while does not draw any DC load current or reduce the gain.

Power Supply Bypassing and Printed Circuit Board Layout

With high phase margin, the EC5565 performs stable gain at high frequency. Like any high-frequency device, good layout of the printed circuit board usually comes with optimum performance. Ground plane construction is highly recommended, lead lengths should be as short as possible and the power supply pins must be well bypassed to reduce the risk of oscillation. For normal single supply operation, where the V_{S^-} pin is connected to ground, a 0.1 μ F ceramic capacitor should be placed from V_{S^+} pin to V_{S^-} pin as a bypassing capacitor. A 4.7 μ F tantalum capacitor should then be connected in parallel, placed in the region of the amplifier. One 4.7 μ F capacitor may be used for multiple devices. This same capacitor combination should be placed at each supply pin to ground if split supplies are to be used.



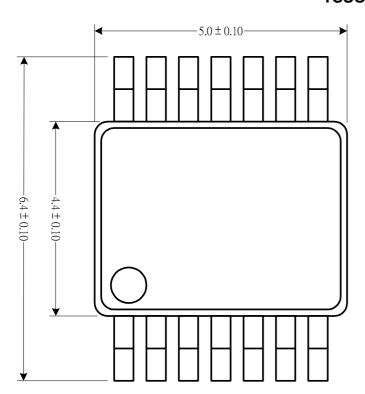
Applications Circuits

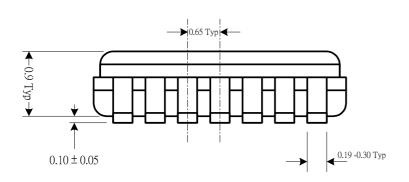


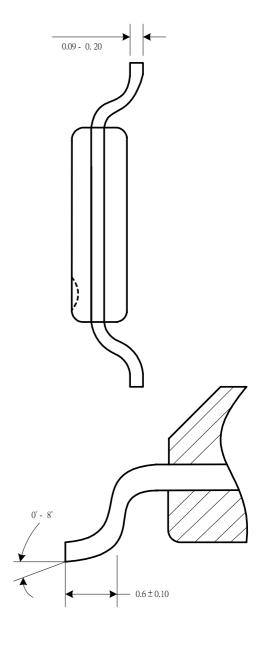


Outline Dimensions (Dimensions shown in millimeters)

TSSOP-14



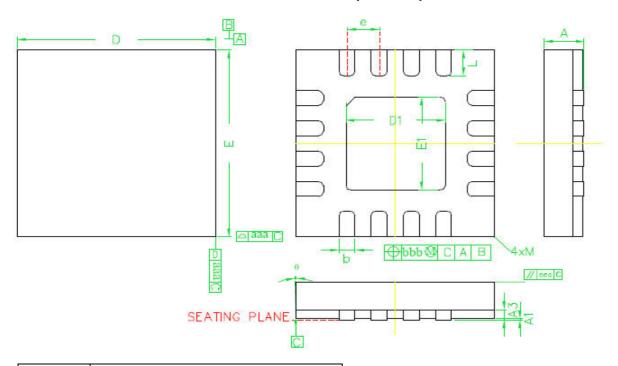






Outline Dimensions

QFN 16L (4x4 mm)



DIMN	Millimeters			
	MIN	NOM	MAX	
Α	0.80	0.90	1.00	
A1	0	0.015	0.030	
А3	_	0.20REF.	_	
b	0.15	0.20	0.25	
D	3.90	4.00BSC	4.10	
D1	2.50	2.60	2.70	
E	3.90	4.00BSC	4.10	
E1	2.50	2.60	2.70	
е	_	0.65BSC	_	
L	0.32	0.40	0.48	
Q	-12	_	0	
aaa	_	0.25	_	
bbb	_	0.10	_	
ccc	_	0.10	_	
М			0.05	
Burr	0	0.030	0.060	

NOTE

- 1. ALL DIMENSIONS ARE IN MILLMETER, Q IS IN DEGREES.
- 2.M: THE MAXIMUM ALLOW ABLE CORNER ON THE MOLDED PLASTIC BODY CORNERS.
- 3.DIMENSION D DOES NOT INCLUDES MOLD PROTRUSIONS OR GATE BURRS.MOLD PROTRUSIONDS AND GATE BURRS SHALL NOT EXCEED 0.15mm PER SIDE.
- 4.DIMENSION E DOES NOT INCLUDES INTERTERMINAL MOLD PROTRUSIONS OR TERMINAL PROTRUSIONS.INTERMINAL MOLD PROTRUSIONS AND/OR TERMINAL PROTRUSIONS SHALL NOT EXCEED 0.20 mm PER SIDE.
- 5. DIMENSION b APPLIES TO PLATED
 TERMINALS.DIMENSION A1 IS PRIMARILY Y TERMINAL
 PLATING, BUT MAY OR MAY NOT INCLUDE A SMALL
 PROTRUSION OF TERMINAL BELOW THE BOTTOM
 SURFACE OF THE PACKAGE.
- 6.JEDEC STANDARD MO-220.